

## Digital Signal Processing

[http://kom.aau.dk/~zt/courses/Digital\\_signal\\_processing/](http://kom.aau.dk/~zt/courses/Digital_signal_processing/)

### Solutions 3 (MM3)

Exercise 3.1. If the sequence  $x[n] = \cos\left(\frac{n\pi}{4}\right), -\infty < n < \infty$

is obtained by sampling the analog (continuous) signal  $x_c(t) = \cos(\Omega_0 t), -\infty < t < \infty$  at a sampling rate of 1000 samples/s, what are two possible values of  $\Omega_0$  that could have resulted in the sequence  $x[n]$  ?

Solution:

$$F_s = 1000 \quad T = 1/1000$$

$$\omega = \Omega_0 T$$

$$\omega = \frac{\pi}{4} \rightarrow \Omega_0 = 250\pi$$

$$\text{In general} \quad \cos\left(\frac{\pi}{4}n\right) = \cos\left(\left(2k\pi + \frac{\pi}{4}\right)n\right)$$

$$\text{choose } \omega = 2\pi + \frac{\pi}{4} \rightarrow \Omega_0 = 2250\pi$$

Exercise 3.2. The input signal,  $x_c(t)$ , given by  $X_c(j\Omega) = 0, |\Omega| \geq 2\pi \times 10^4$

is subjected to a sampling system, using sampling period  $T$ , and having output  $x[n]$ :

$$x[n] = x_c(nT_s).$$

The signal  $x[n]$  is then subjected to a digital filter, having impulse response  $h[n]$ , and

generating the output  $y[n] = T \sum_{k=-\infty}^n x[k]$ .

a) What is the maximum allowable value of  $T$  if aliasing is to be avoided ? (so that  $x_c$  can be recovered from  $x[n]$ ) ?

b) What is  $h[n]$  ?

Solution:

$$a) F_s > 2F_0 \rightarrow T_s < \frac{1}{2 \cdot 10^4}$$

b)

$$y[n] = T_s \sum_{k=-\infty}^n x[k]$$

$$h[n] = T_s \sum_{k=-\infty}^n \delta[k] = T_s \cdot u[n]$$

Exercise 3.3 Consider the sampling and reconstruction system given below where  $x(t)$  is given by the formula  $x(t) = 10 \cos(20\pi t - \pi/4) - 5 \cos(50\pi t)$

a) What condition must be satisfied by the sampling rate to ensure  $y(t) = x(t)$ ?

b) How should  $f_s$  be chosen, so that  $y(t) = A + 10 \cos(20\pi t - \pi/4)$ ?

c) What is the value of the constant A?

Solution:

a)

$$F_0 = 25 \text{ Hz}$$

$$\therefore F_s > 50 \text{ Hz} \quad \text{to avoid aliasing}$$

b)

In  $y(t)$ , the 25 Hz component is aliased to 0 Hz. This is realised by choosing  $F_s = 25 \text{ Hz}$  which at the same time assures that no aliasing for the 10 Hz component  $10 \cos(20\pi t - \pi/4)$ .

c)

$$x[n] = 10 \cos(20\pi n / 25 - \pi/4) - 5 \cos(50\pi n / 25) = 10 \cos(20\pi n / 25 - \pi/4) - 5$$

Reconstruction of the signal :

$$y(t) = 10 \cos(20\pi t - \pi/4) - 5$$

$$\therefore A = -5$$

Exercise 3.4. Consider a stable LTI system with input  $x[n]$  and output  $y[n]$ . The input and output satisfy the difference equation

$$y[n-1] - \frac{10}{3} y[n] + y[n+1] = x[n]$$

- a) Find the system function  
 b) Plot the poles and zeros in the z-plane.

Solution:

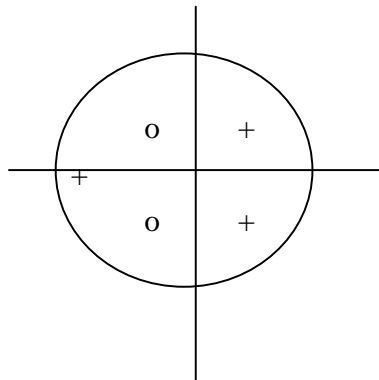
a)

$$H(z) = \frac{z^{-1}}{z^{-2} - \frac{10}{3}z^{-1} + 1} = \frac{z^{-1}}{(z^{-1} - 3)(z^{-1} - \frac{1}{3})} = \frac{z}{(1 - 3z)(1 - \frac{1}{3}z)}$$

b)

Poles :  $z = 3, \frac{1}{3}$ ; zeros :  $z = 0, \infty$

Exercise 3.5. If the system function  $H(z)$  of a LTI system has a pole-zero diagram as shown in the following figure and the system is causal, can the inverse system  $H_i(z)$ , where  $H(z)H_i(z)=1$ , be both causal and stable? Clearly justify your answer.



Solution:

$H_i(z)$  cannot be both causal and stable since the zero of  $H(z)$  at  $z = \infty$  is a pole of  $H_i(z)$  and the existence of a pole at  $z = \infty$  implies that the system is not causal.